Time-Frequency Distributions for Non-Stationary Signal Analysis

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PHD Days ONERA PHY - 15/01/2020



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Fourier Analy	rsis		

- Fourier Analysis is based on the universality of the concept of frequency:
 - physical waves: acoustic, vibrations, geophysic, optic, electromagnetism, etc.
 - periodicity of events: economy, biology, astronomy, etc.
- Its mathematical structure which naturally lends itself to common transformations such as linear filtering (complex exponential waves are eigenvectors of the convolution operator) by translating them in a very simple way.



• Fourier analysis has led to the development of numerous algorithms, processors and devices for frequency analysis.

Fourier Transform: Signal decomposition onto eternal pure monochromatic waves basis

$$x(t) = \int_{-\infty}^{+\infty} X(f) \exp(2i\pi f t) df \text{ where } X(f) = \int_{-\infty}^{+\infty} x(t) \exp(-2i\pi f t) dt$$

Fourier analysis is well well adapted for **stationary signals**: signal whose spectral components do not vary with time (pure monochromatic waves, linear combination of pure monochromatic waves)



Non-Stationary Signals: signals whose spectral components vary with time

- Most of the information is carried by the non-stationarities: starting, end, transients
- Spectral analysis loses temporal information since it averages over time.
- Local analysis approach: adapt the stationary tools (TF, linear model) to the variations over time.



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Some Solution	s for Non-Sta	tionary Signals Anal	ysis	

A solution: Parametric methods on time slices of the signal.

Examples of parametric spectral methods: AR, ARMA, MUSIC:

- Adaptive/evolving methods. Problem: need to estimate the order of the model on each local time slices
- A priori knowledge difficult to exploit on non-stationarities.

Alternative: Time-Frequency Analysis.

Global analysis of time-dependent spectral characteristics:

- The problem of estimating the model order is avoided. Problem: choosing the type of analysis. The analysis can play the role of a "model".
- Difficulty of interpretation of the analysis (uncertainty principle, interferences).



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First (simple) idea: define a local frequency (which depends on time)

- Write the signal in terms of modulated amplitude and phase (positive frequency analytical signal): $x(t) = a(t) \exp(i\phi(t))$.
- Phase variations define the so-called instantaneous frequency: $f_x(t) = \frac{1}{2\pi} \frac{\partial \phi}{\partial t}$. Examples
 - Monochromatic wave: the instantaneous frequency of the signal $x(t) = \cos(2\pi f_0 t)$ is $f_x(t) = f_0$ (constant).
 - Chirp (linearly frequency modulated signal): the instantaneous frequency of the signal $x(t) = \cos\left(2\pi\left(2 a t^2 + b t + c\right)\right)$ is $f_x(t) = a t + b$

Dual notion: Group Delay

the frequency variations of phase of $X(f) = A(f) \exp(i\psi(f))$ define the group delay (time dependent on frequency):

$$t_{x}(f) = -rac{1}{2 \, \pi} rac{\partial \psi}{\partial f}$$

Example: the signal $x(t) = \delta(t - t_0)$ as the group delay $t_x(f) = t_0$.





Linear Time-Frequency Analysis: Signal decomposition onto a dedicated basis





Heisenberg Inequality: For all signal window h(.) and defining the spectral bandwidth σ_f^h and the time duration σ_t^h , we have:

$$\sigma_{f}^{h} \sigma_{t}^{h} \geq 1$$

The Gaussian shape signal is the only signal verifying the above equality $(\sigma_f^h \sigma_t^h = 1)$.



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Readability, access to instantaneous frequency or group delay of signals, robust to noise



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- Theory of operators: mean value of operator (observable) like in Quantum Mechanics Theory.
- Theory of multivariate probability density functions, Theory of copulas.

Goal: Designing a joint time-frequency distribution function P(t, f) which

• has to distribute the total energy E_z of the signal z(.) in the whole Time-Frequency plane:

$$\int_{-\infty}^{+\infty}\int_{-\infty}^{+\infty}P(t,f)\,dt\,df=E_{z}=\int_{-\infty}^{+\infty}\left|z(t)
ight|^{2}\,dt=\int_{-\infty}^{+\infty}\left|Z(f)
ight|^{2}\,df$$

• has to retrieve the following marginal distribution functions:

$$\int_{-\infty}^{+\infty} P(t,f) dt = |Z(f)|^2 \quad \text{and} \quad \int_{-\infty}^{+\infty} P(t,f) df = |z(t)|^2$$



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Required Prop	perties 1/2		

• Covariance Principle by action of a group: Consistency of the distribution P_z after a physical transformation group \mathcal{T} applied on the signal z(.) and on its dual representation $\mathcal{T}'P_z$

$$\begin{array}{cccc} z & \longrightarrow & P_z \\ \mathcal{T} & & \mathcal{T}' \\ \mathcal{T} z & \longrightarrow & \mathcal{T}' P_z = P_{\mathcal{T} z} \end{array}$$

Examples: time translation, frequency translation (approximate Doppler effect), scaling (true Doppler effect), rotation, Lorentz, etc.

• **Unitarity** (Conservation of the inner product, Moyal Formula, Extended Parseval Theorem):

$$\int_{-\infty}^{+\infty} \int_{-\infty}^{+\infty} P_{z_1}(t,f) P_{z_2}(t,f) dt df = \left| \int_{-\infty}^{+\infty} z_1(t) z_2^*(t) dt \right|^2$$

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Required Pro	perties 2/2		

• Access to the instantaneous frequency $f_z(.)$ and group delay $t_z(.)$ through marginalization:

$$f_{z}(t) = \frac{1}{2\pi} \frac{\partial \phi}{\partial t} = \frac{\int_{-\infty}^{+\infty} f P(t, f) df}{\int_{-\infty}^{+\infty} P(t, f) df} \quad \text{and} \quad t_{z}(f) = -\frac{1}{2\pi} \frac{\partial \psi}{\partial f} = \frac{\int_{-\infty}^{+\infty} t P(t, f) dt}{\int_{-\infty}^{+\infty} P(t, f) dt},$$

- Positivity,
- Conservation of spectral and time signal supports,
- Perfect time-frequency localization on some particular signals:
 - Monochromatic signals: $z(t) = \exp(2i \pi f_0 t) \longrightarrow P_z(t, f) = \delta(f f_0)$,
 - Shock type signals: $z(t) = \delta(t t_0) \longrightarrow P_z(t, f) = \delta(t t_0)$,
 - Monocomponent frequency modulated signals:

$$z(t) = \exp(i\phi(t)) \longrightarrow P_z(t, f) = \delta(f - f_z(t)) = \delta\left(t - \frac{1}{2\pi}\frac{\partial\phi}{\partial t}\right)$$



Introduction to Spectral Analysis OCOO How to Build the Quadratic Time-Frequency Distributions

Building through kernel type model respecting as many required properties as possible

In order to preserve the primary notion of energy distribution in the time-frequency plane, these distributions are constructed on bilinear forms $z(u)z^*(v)$ of the time signal (or, equivalently, frequency $Z(u)Z^*(v)$):

$$P(t,f) = \int_{-\infty}^{+\infty} \int_{-\infty}^{+\infty} K(u,v;t,f) z(u) z^{*}(v) du dv$$

=
$$\int_{-\infty}^{+\infty} \int_{-\infty}^{+\infty} \hat{K}(u,v;t,f) Z(u) Z^{*}(v) du dv$$

The kernel K(u, v; t, f) (or $\hat{K}(u, v; t, f)$) has to be determined according to the required properties.



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Wevl-Heisenb	erg group and	the Cohen's class	

Determination of the class of covariant solutions through the time and frequency translation transformation group (Weyl-Heisenberg):

$$\begin{array}{cccc} z(t) & \longrightarrow & P_z(t,f) \\ \mathcal{T} \downarrow & & \mathcal{T}' \downarrow \\ z'(t) = z(t-t_0) \exp\left(2i\pi f_0 t\right) & \longrightarrow & P_{z'}(t,f) = P_z(t-t_0,f-f_0) \end{array}$$

Direct solutions: Cohen's class parametrized by a 2D kernel $\psi(.,.)$ (or equivalently $\Pi(.,.)$):

$$P_{z}(t,f) = \int_{-\infty}^{+\infty} \int_{-\infty}^{+\infty} \int_{-\infty}^{+\infty} e^{2i\pi\xi(s-t)} \psi(\xi,\tau) z\left(s+\frac{\tau}{2}\right) z^{*}\left(s-\frac{\tau}{2}\right) e^{-2i\pi f\tau} d\xi \, ds \, d\tau$$
$$= \int_{-\infty}^{+\infty} \int_{-\infty}^{+\infty} \Pi(s-t,\xi-f) W_{z}(s,\xi) \, ds \, d\xi$$

where the central distribution W_z is the well known Wigner-Ville distribution:

$$W_z(t,f) = \int_{-\infty}^{+\infty} z\left(t+\frac{\tau}{2}\right) \, z^*\left(t-\frac{\tau}{2}\right) \, e^{-2i\pi f \tau} \, d\tau$$



Time Frequency Analysis

Cohen's Class and Weyl-Heisenberg Group

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Bertrand's Class and Affine Group

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Some distributions of the Cohen's class

Nom	$\psi(\xi,\tau)$	$P_z(t,f)$
Wigner-Ville	1	$\int_{-\infty}^{+\infty} z\left(t+\frac{\tau}{2}\right) z^*\left(t-\frac{\tau}{2}\right) e^{-2i\pi f\tau} d\tau$
s-Wigner	$e^{2i\pi s\xi\tau}$	$\int_{-\infty}^{+\infty} z \left(t - \left(s - \frac{1}{2} \right) \tau \right) z^* \left(t - \left(s + \frac{1}{2} \right) \tau \right) e^{-2i\pi f \tau} d\tau$
Rihaczek	$e^{i\pi\xi\tau}$	$z(t) Z^*(f) e^{-2i\pi f au}$
Born-Jordan	$\frac{\sin \pi \xi \tau}{\pi \xi \tau}$	$\int_{-\infty}^{+\infty} \left[\frac{1}{ \tau } \int_{t- \tau /2}^{t+ \tau /2} z\left(s+\frac{\tau}{2}\right) z^*\left(s-\frac{\tau}{2}\right) ds \right] e^{-2i\pi f \tau} d\tau$
Choï-Williams	$e^{-(\pi\xi\tau/\sigma)^2/2}$	$\int_{-\infty}^{+\infty} \int_{-\infty}^{+\infty} \frac{\sigma}{ \tau } e^{-2\sigma^2(s-t)^2/\tau^2} z\left(s+\frac{\tau}{2}\right) z^*\left(s-\frac{\tau}{2}\right) e^{-2i\pi f\tau} ds d\tau$
Spectrogramme	$A_h(\xi, \tau)$	$\left \int_{-\infty}^{+\infty} z(s) h^*(s-t) e^{-2i\pi f s} ds\right ^2$
Séparable	$G(\xi) h(\tau)$	$\int_{-\infty}^{+\infty} \int_{-\infty}^{+\infty} h(\tau) g(s-t) z\left(s+\frac{\tau}{2}\right) z^*\left(s-\frac{\tau}{2}\right) e^{-2i\pi f\tau} ds d\tau$

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• Wigner-Ville [E. P. Wigner, 1932]: main and most popular distribution:

$$W_z(t,f) = \int_{-\infty}^{+\infty} z\left(t+\frac{\tau}{2}\right) z^*\left(t-\frac{\tau}{2}\right) e^{-2i\pi f\tau} d\tau = \int_{-\infty}^{+\infty} Z\left(f+\frac{\nu}{2}\right) Z^*\left(f-\frac{\nu}{2}\right) e^{-2i\pi t\nu} d\nu$$



• Link to Woodward's Radar Ambiguity Function $A_z(\tau, \nu)$ through Fourier Transform

$$A_z(\tau,\nu) = \int_{-\infty}^{+\infty} z\left(u+\frac{\tau}{2}\right) z^*\left(u-\frac{\tau}{2}\right) e^{-2i\pi\nu u} du = \int_{-\infty}^{+\infty} W_z(t,f) e^{-2i\pi(t\nu+f\tau)} dt df$$



- No distribution satisfies all the constraints (e.g. positivity and unity).
- Illegibility due to interference between elementary signal components
- Need to erase or reduce interferences between each time and spectral components







• Regularized version of the Wigner-Ville distribution: smoothing operation to reduce interferences between components (ex: Short Time Fourier Transform, Gabor)

$$P_{l}(t_{0}, f_{0}) = \int_{-\infty}^{+\infty} \int_{-\infty}^{+\infty} P_{z}(t, f) P_{h}(t - t_{0}, f - f_{0}) dt df = \left| \int_{-\infty}^{+\infty} z(u) h^{*}(u - t_{0}) e^{-2i\pi f_{0} u} du \right|^{2}$$

where $h(t) \exp(2i\pi f_0 t)$ is an analyzing window localized around t = 0 and $f = f_0$.



• Drawback: Heisenberg inequality, no control on joint time and spectral resolutions







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Going further	with Affine (Group		

Determination of the class of covariant solutions through the time-scale and time-shift transformation group (Affine group):

Direct solutions: Bertrand's class parametrized by a 2D kernel K(u, v):

$$P_Z(t,f) = \int_0^{+\infty} \int_0^{+\infty} K(u,v) Z(f u) Z^*(f v) e^{2i\pi f t (u-v)} du dv$$

Example of the central distribution: Bertrand's Unitary Affine Distribution:

$$P_{Z}^{0}(t,f) = f^{2r+2-q} \int_{-\infty}^{+\infty} \left[\frac{u}{2\sinh(u/2)} \right]^{2r+2} Z\left(f \frac{u e^{u/2}}{2\sinh(u/2)} \right) Z^{*}\left(f \frac{u e^{-u/2}}{2\sinh(u/2)} \right) e^{2i\pi f t u} du$$



 Regularized version of the Unitary Affine distribution: smoothing operation to reduce interferences between components (Wavelet Transform)

$$\tilde{P}(t_0, f_0) = \int_{-\infty}^{+\infty} \int_0^{+\infty} P_Z^0(t, f) P_{\Psi}^0\left(\frac{f}{f_0}(t - t_0), \frac{f_0}{f}\right) dt df = \left|\int_{-\infty}^{+\infty} Z(f) \Psi^*\left(\frac{f}{f_0}\right) e^{-2i\pi f t_0} df\right|^2$$

where $\Psi(f) \exp(2i\pi f t)$ is an analyzing mother window localized around t = 0 and f = 1.



• Drawback: Heisenberg inequality, no control on joint time and spectral resolutions

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150 Time fal

150 Time (a)

10

150 Time (s)





F. Boitier, A. Godard, N. Dubreuil, P. Delaye, C. Fabre, and E. Rosencher, Nature Communications 2, 425 (2011).













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- Conventional Fourier Imaging (laboratory, SAR, ISAR) does not exploit the potential non-stationarities of the scatterers
- Hypothesis of bright points modeling: all the scatterers localized in **x** and characterized by the complex spatial amplitude distribution $I(\mathbf{x})$ have **the same behavior** for any wave vector $\mathbf{k} = \frac{2f}{c} (\cos \theta, \sin \theta)^T$. After some processing, the backscattering coefficient $H(\mathbf{k})$ acquired by the radar is simply related to the SAR image $I(\mathbf{x})$ through:

$$H(\mathbf{k}) = \int_{\mathcal{D}_{\mathbf{x}}} I(\mathbf{x}) \exp\left(-2 i \pi \mathbf{k}^{T} \mathbf{x}\right) d\mathbf{x}$$



• The SAR image $I(\mathbf{x})$ is then obtained through the Inverse Fourier Transform: $I(\mathbf{x}) = \int_{\mathcal{D}_{\mathbf{k}}} H(\mathbf{k}) \exp(2i\pi \mathbf{k}^T \mathbf{x}) d\mathbf{k}$

With this model, all information relative to frequency f and angle θ are lost. Hence, spectral and angular diversities are lost.



elevation 30°

elevation 50°



Sub-band 1 Sub-band 2 Sub-band 3

Scatterers have different behaviors with respect to the frequency and illumination direction.



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Applications:	Radar Imagir	ıg	

Time-Frequency Distributions are generally devoted to non-stationary time signals analysis. They can be easily extended in 2D.

Key idea: In the context of SAR Imaging, Time-Frequency Analysis allows:

- to highlight the coloration and anisotropy properties of monodimensional SAR scatterers,
- to characterize each pixel of the complex SAR image with a vector of information related to angular or/and frequency behaviors.

Time Frequeny Analysis and the physical group theory (Heisenberg or affine group) allow to construct **hyperimages** [Bertrand 91, Ovarlez 92] through:

$$\tilde{l}(\mathbf{r}_{0},\mathbf{k}_{0}) = < H(.), \Psi_{\mathbf{r}_{0},\mathbf{k}_{0}}(.) > = \int_{\mathcal{D}_{\mathbf{k}}} H(\mathbf{k}) \, \Psi^{*}_{\mathbf{r}_{0},\mathbf{k}_{0}}(\mathbf{k}) \, d\mathbf{k} \, ,$$

where $\Psi_{r_0,k_0}(\mathbf{k})$ is a family of wavelet bases (Gabor, wavelet) generated from a mother wavelet $\phi(f,\theta)$ through the chosen physical group of transformation (translations, scale in frequency, etc.) and where $\mathcal{D}_{\mathbf{k}}$ is the spectral/angular support of Ψ .







Anechoic room geometry

















From Mono-Channel to Multi-Channel SAR Image: Example of $N_f = 3$ sub-bands and $N_{\theta} = 3$ sub-looks image decomposition:



Exploitation of the diversity for target detection and change detection

Each pixel *i* of the mono-channel SAR image can now be characterized by a *N*-vector $\mathbf{x}_i = \begin{bmatrix} W_{1,1}^i, \dots, W_{N_f,N_\theta}^i \end{bmatrix}^T$ of information $(N = N_f N_\theta)$ related to **dispersion** in frequency domain and **anisotropy** in angular domain [P. Formont 13, A. Mian 19].