

Off-grid Radar Target Detection with the Normalized Matched Filter: a Monopulse-Based Detection Scheme

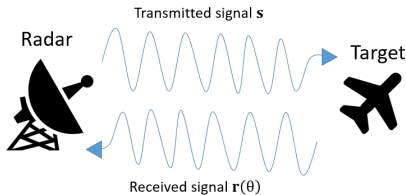
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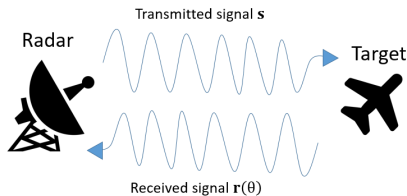
Context: The Radar detection problem

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- Emit signal, and search for echoes in received signal.



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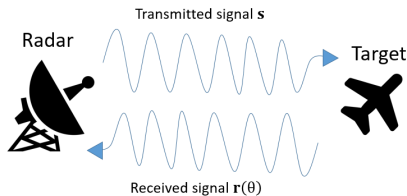
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- Emit signal, and search for echoes in received signal.



- Received signal depends on unknown target parameters θ .
- For practical reasons, tests are run for fixed values of parameters θ_0 in a Grid $G = \{k\Delta, k \in [0 .. N - 1]\}$, with N the number of samples and Δ the sampling interval. Cell : $[\theta_0 - \Delta/2, \theta_0 + \Delta/2]$.

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- In real conditions, there is no reason to have $\theta = \theta_0$. We have mismatch : $\theta \neq \theta_0$, and performance derived under on-grid model is not met.
- This motivates the search of a robust detection scheme.

- 1 Problem formulation
 - Model under study
 - GLRT
 - Off-Grid
- 2 A monopulse-based solution
 - Definitions
 - The Procedure
 - Properties
- 3 Numerical Results
 - Detector Comparison
 - Simulation under white noise
- 4 Bibliography

The Radar detection problem

The classical Radar detection problem is the following binary Hypothesis Test:

$$\begin{cases} H_0 : \mathbf{r} = \mathbf{n} \\ H_1 : \mathbf{r} = \alpha s(\theta) + \mathbf{n} \end{cases}, \text{ where}$$

- $\mathbf{r} \in \mathbb{C}^N$ is the observation,
- $s(\theta) \in \mathbb{C}^N$ is the signal echo reflected by a target with parameters θ (range, angle, Doppler...),
- $\alpha \in \mathbb{C}$ is the complex amplitude of the received signal,
- $\mathbf{n} \in \mathbb{C}^N$ is the additive noise vector, independent of the source signal.
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Here the signal $s(\theta)$ follows the general spectral analysis model (angle or Doppler shift in Radar):

$$s(\theta) = \frac{1}{\sqrt{N}} \left[1, e^{2i\pi\theta}, \dots, e^{2i\pi(N-1)\theta} \right]^T.$$

with $\Delta = 1/N$: grid vectors are orthogonal.

The Generalized Likelihood Ratio Test

The GLRT is:

$$\Lambda(\mathbf{r}) = \frac{\max_{\lambda_1} f_{H_1}(\mathbf{r})}{\max_{\lambda_0} f_{H_0}(\mathbf{r})} \underset{H_0}{\overset{H_1}{\geq}} \eta.$$

where

- for $i \in \{0, 1\}$, f_{H_i} is the density function of \mathbf{r} under H_i and λ_i are the unknown parameters under H_i ,
- η guarantees a fixed Probability of False Alarm (PFA).

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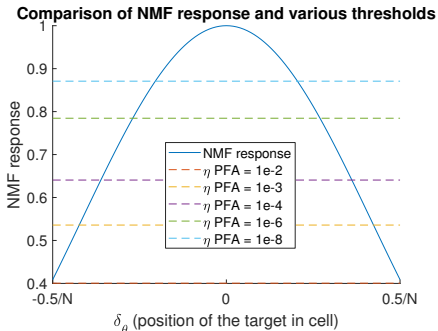
When $\lambda_1 = \{\sigma, \alpha\}$ and $\lambda_0 = \{\sigma\}$, with θ known, the GLRT is the following Normalized Matched Filter (NMF)

[Scharf and Lytle, 1971]:

$$t_{\Gamma}(\mathbf{r}, \theta) = \frac{\left| \mathbf{s}(\theta)^H \boldsymbol{\Gamma}^{-1} \mathbf{r} \right|^2}{\left(\mathbf{s}(\theta)^H \boldsymbol{\Gamma}^{-1} \mathbf{s}(\theta) \right) \left(\mathbf{r}^H \boldsymbol{\Gamma}^{-1} \mathbf{r} \right)} \underset{H_0}{\overset{H_1}{\geq}} \eta.$$

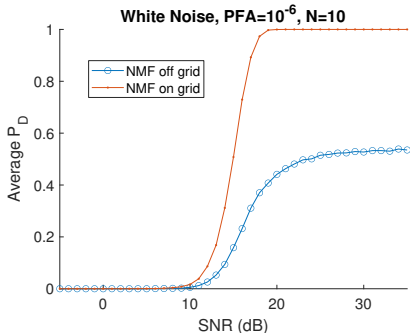
Impact of the off-grid target on NMF

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Impact of the off-grid target on NMF

- Mismatch $\delta = \theta - \theta_0$
- Angle mismatch creates a degradation of the NMF response even without noise
- When θ uniformly distributed in a cell it can be shown $P_D \rightarrow 1$
[Rabaste et al., 2016]
- Even worse when $\Gamma \neq \mathbf{I}$



- Extension of the GLRT to off-grid targets:

$$\text{GLRT}(\mathbf{r}, \theta_0) = \max_{\theta_c \in [\theta_0 - \Delta/2, \theta_0 + \Delta/2]} t_{\Gamma}(\mathbf{r}, \theta_c) \underset{H_0}{\overset{H_1}{\geq}} \eta.$$

The best P_D , no closed form available, threshold unknown, precise approximation can be costly.

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- Existing sub-optimal cost-efficient solutions include
 - Oversampling approximate GLRT, threshold unknown
 - Using DPSS subspace to approximate the cell structure, analytical threshold [Bosse and Rabaste, 2018]
 - Detection with bounded mismatch, not yet suited to low PFA Radar context [Besson, 2006]

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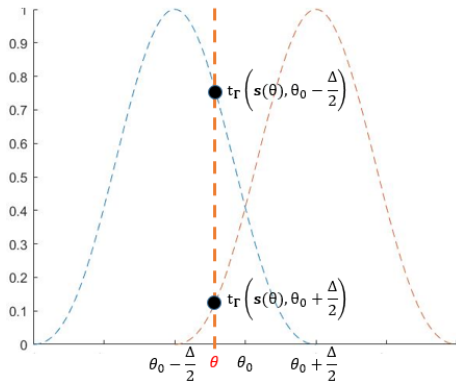
$$\text{GLRT}(\mathbf{r}, \theta_0) = \max_{\theta_c \in [\theta_0 - \Delta/2, \theta_0 + \Delta/2]} t_{\Gamma}(\mathbf{r}, \theta_c) \underset{H_0}{\overset{H_1}{\geq}} \eta.$$

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- These solutions do not correct the convergence issue for all Γ and are not always near GLRT.

Proposed Monopulse Inspired Scheme

- Monopulse traditionally used to estimate target parameters from a single pulse [Mosca, 1969].
- The idea is to combine two tests in a function h that carries info about θ .
- Used with noise, h can give an approximation $\hat{\delta}$



Monopulse Functions

- Classically in monopulse, the function h is:

$$h_{\mathbf{r},\theta_0}(\mathbf{r}) = \frac{t_{\mathbf{r}}\left(\mathbf{r}, \theta_0 - \frac{\Delta}{2}\right) - t_{\mathbf{r}}\left(\mathbf{r}, \theta_0 + \frac{\Delta}{2}\right)}{t_{\mathbf{r}}\left(\mathbf{r}, \theta_0 - \frac{\Delta}{2}\right) + t_{\mathbf{r}}\left(\mathbf{r}, \theta_0 + \frac{\Delta}{2}\right)}.$$

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- mismatch: $\delta = \theta - \theta_0$, noise-free function g :

$$g_{\mathbf{r},\theta_0}(\delta) = h_{\mathbf{r},\theta_0}(s(\theta_0 + \delta))$$

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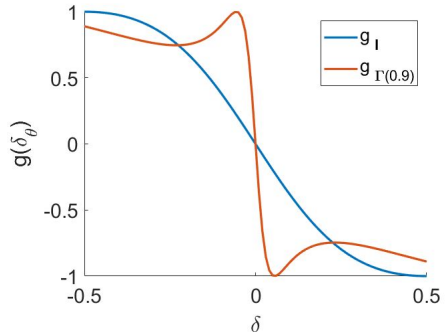
$$g_{\mathbf{r},\theta_0}(\delta) = h_{\mathbf{r},\theta_0}(s(\theta_0 + \delta))$$

- Goal: compute $\hat{\delta}$ by inverting $g(\delta)$ thanks to h applied on noisy signal \mathbf{r} .

Choice of g

$$\Gamma(\rho) = \mathcal{T} \left(\begin{bmatrix} 1 & \rho & \dots & \rho^{N-1} \end{bmatrix} \right)$$

- g_{Γ, θ_0} needs to be invertible. This is not always the case.

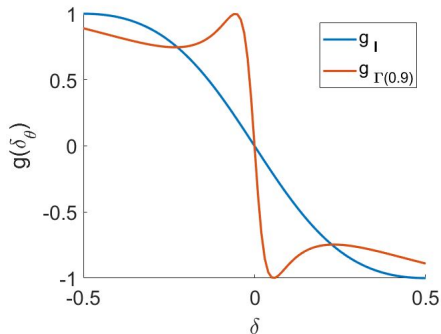


Candidate $g(.)$ functions for
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- g_{Γ, θ_0} needs to be invertible. This is not always the case.
- We use g_I in the following even with colored Gaussian noise. We note it g .



Candidate $g(.)$ functions for
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The Procedure

- The test procedure is the following, for every θ_0 of the grid:

The test procedure

- 1 compute $t_I \left(\mathbf{r}, \theta_0 - \frac{\Delta}{2} \right)$ and $t_I \left(\mathbf{r}, \theta_0 + \frac{\Delta}{2} \right)$;
- 2 compute $\hat{\delta} = g^{-1} (h_{I, \theta_0}(\mathbf{r}))$;
- 3 run the final tests $t_r(\mathbf{r}, \hat{\delta} + \theta_0) \underset{H_0}{\overset{H_1}{\geq}} \eta_g$.

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- the statistic of $t_r(\mathbf{r}, \hat{\delta} + \theta_0)$ depends on the non-independent random variables \mathbf{r} and $\hat{\delta} \implies$ no closed form available for η_g
- η_g is approximated with Monte Carlo simulations

Properties of this approach

Let us describe some properties of this approach:

- Only $2N$ tests are run for the whole spectral space, and the rest of the computations are simply lookup table operations,
- When the SNR tends to infinity, $\hat{\theta} = \theta$ and the Probability of Detection (PD) tends to 1,
- When $\Gamma = \mathbf{I}$, $\hat{\theta}$ is an approximate MLE and our test is an approximate GLRT [[Mosca, 1969](#)],
- When $\Gamma \neq \mathbf{I}$, our test is still close to the GLRT in term of performance.

Detector Comparison

We compare our scheme to detectors of similar cost:

- An oversampled NMF with 2 tests per cell,
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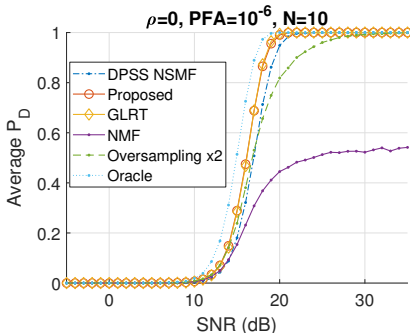
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We also compare it to :

- The classical NMF
- An approximate GLRT using 50 tests per cell
- The Oracle detector, which knows where the target is and as such is the best detector possible

Simulation under white noise

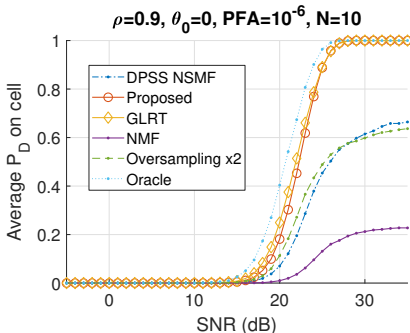
- Target parameter θ drawn at random uniformly.
- Our detector converges to 1 asymptotically and outperforms other detectors in the same computational range.



P_D of the detectors under white noise, for a P_{FA} of 10^{-6} , $N = 10$.

Simulation under colored noise

- Target parameter θ drawn at random uniformly in $[\theta_0 - \Delta/2, \theta_0 + \Delta/2]$.
- In this cell, our detector stays close to the GLRT and does greatly better than the other detectors, which do not converge to 1.



P_D of the detectors with $\rho = 0.9$,
 $\theta_0 = 0$ for a P_{FA} of 10^{-6} , $N = 10$.

We introduced a new detector that approximates GLRT under white noise for off-grid targets.

- simply based on the well-known monopulse procedure, classically used in array processing.

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- simply based on the well-known monopulse procedure, classically used in array processing.
- performance is close to GLRT while being cost-efficient
- Future works will investigate the performance of our detector under adaptive context, other noise models, and PFA-threshold relationship.

The End

Thank You For Listening !

- [Bandiera et al., 2009] Bandiera, F., Orlando, D., and Ricci, G. (2009). *Advanced Radar Detection Schemes Under Mismatched Signal Models*. Morgan & Claypool publishers.
- [Besson, 2006] Besson, O. (2006).
Detection of a signal in linear subspace with bounded mismatch.
Aerospace and Electronic Systems, IEEE Transactions on, 42(3):1131–1139.
- [Bosse and Rabaste, 2018] Bosse, J. and Rabaste, O. (2018).
Subspace rejection for matching pursuit in the presence of unresolved targets.
Signal Processing, IEEE Transactions on, 66(8):1997–2010.
- [Bosse et al., 2020] Bosse, J., Rabaste, O., and Ovarlez, J.-P. (2020).
Adaptive subspace detectors for off-grid mismatched targets.
ICASSP 2020-2020 IEEE International Conference on Acoustics, Speech and Signal Processing (ICASSP), pages 4777–4780.
- [Chaumette, 2004] Chaumette, E. (2004).
Contribution à la caractérisation des performances des problèmes conjoints de détection et d'estimation.
PhD thesis, Cachan, Ecole Normale Supérieure, Gif-sur-Yvette, France.
- [Ciunzio et al., 2016] Ciunzio, D., De Maio, A., and Orlando, D. (2016).
A unifying framework for adaptive radar detection in homogeneous plus structured interference - part II: Detectors design.
Signal Processing, IEEE Transactions on, 64:2907–2919.
- [Conte et al., 1995] Conte, E., Lops, M., and Ricci, G. (1995).
Asymptotically optimum radar detection in compound-Gaussian clutter.
Aerospace and Electronic Systems, IEEE Transactions on, 31(2):617–625.
- [Mosca, 1969] Mosca, E. (1969).
Angle estimation in amplitude comparison monopulse systems.
Aerospace and Electronic Systems, IEEE Transactions on, AES-5(2):205–212.
- [Ollila et al., 2012] Ollila, E., Tyler, D. E., Koivunen, V., and Poor, H. V. (2012).
Complex elliptically symmetric distributions: Survey, new results and applications.

- [Pascal et al., 2006] Pascal, F., Ovarlez, J.-P., Forster, P., and Larzabal, P. (2006).
On a SIRV-CFAR detector with radar experimentations in impulsive noise.
In *European Signal Processing Conference, EUSIPCO'06*, Florence, Italy.
- [Rabaste et al., 2016] Rabaste, O., Bosse, J., and Ovarlez, J.-P. (2016).
Off-grid target detection with Normalized Matched Subspace Filter.
In *24th European Signal Processing Conference (EUSIPCO)*, pages 1926–1930.
- [Rabaste and Trouve, 2014] Rabaste, O. and Trouve, N. (2014).
Geometrical design of radar detectors in moderately impulsive noise.
Aerospace and Electronic Systems, IEEE Transactions on, 50(3):1938–1954.
- [Scharf and Friedlander, 1994] Scharf, L. L. and Friedlander, B. (1994).
Matched subspace detectors.
Signal Processing, IEEE Transactions on, 42(8):2146–2157.
- [Scharf and Lytle, 1971] Scharf, L. L. and Lytle, D. W. (1971).
Signal detection in Gaussian noise of unknown level: an invariance application.
Information Theory, IEEE Transactions on, 17:404–411.