

On the false alarm probability of the Normalized Matched Filter for off-grid target detection IEEE ICASSP 2022

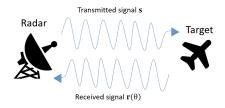
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April 5, 2022

Context: The Radar detection problem

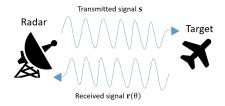
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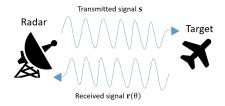


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Formalization: The Radar detection problem

The classical Radar detection problem is the following binary Hypothesis Test:

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\left\{ \begin{array}{l} H_0: {\bf r}={\bf n} \\ H_1: {\bf r}=\alpha\, {\bf s}(\theta)+{\bf n} \end{array} \right. , \text{ where }
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- $\mathbf{r} \in \mathbb{C}^{N}$ is the observation,
- $s(\theta) \in \mathbb{C}^N$ is the signal echo reflected by a target with parameters θ (range, angle, Doppler...),
- $\alpha \in \mathbb{C}$ is the complex amplitude of the received signal,
- $\mathbf{n} \in \mathbb{C}^{N}$ is the additive noise vector, independent of the source signal. Our results hold for any spherically invariant distribution such as $\mathbf{n} \sim \mathcal{CN}(\mathbf{0}, \sigma^{2} \Gamma)$.



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- $\mathbf{s}(\theta)$: General spectral analysis model (angle or Doppler with Radar) :

$$\mathbf{s}(\boldsymbol{\theta}) = \frac{1}{\sqrt{N}} \left[1, e^{2i\pi\boldsymbol{\theta}}, \dots, e^{2i\pi(N-1)\boldsymbol{\theta}} \right]^{\mathsf{T}}$$



The Generalized Likelihood Ratio Test is given by:

$$\Lambda(\mathbf{r}) = \frac{\displaystyle\max_{\lambda_1} f_{H_1}(\mathbf{r})}{\displaystyle\max_{\lambda_0} f_{H_0}(\mathbf{r})} \mathop{\gtrless}_{H_0}^{H_1} \eta.$$

where

- for $i \in \{0, 1\}$, f_{H_i} is the density function of \mathbf{r} under H_i and λ_i are the unknown parameters under H_i ,
- η guarantees a fixed Probability of False Alarm (PFA).



The Normalized Matched Filter

• When $\lambda_1 = \{\sigma, \alpha\}$ and $\lambda_0 = \{\sigma\}$, with θ known, the GLRT reduces to the following Normalized Matched Filter (NMF) [Scharf and Lytle, 1971]: $|s(\theta)^{H} \Gamma^{-1} r|^{2}$ H₁

$$t_{\Gamma}(\mathbf{r},\theta) = \frac{|\mathbf{s}(\theta)^{-1} \mathbf{r}^{-1}|}{\left(\mathbf{s}(\theta)^{H} \Gamma^{-1} \mathbf{s}(\theta)\right) \left(\mathbf{r}^{H} \Gamma^{-1} \mathbf{r}\right)} \overset{\text{def}}{\underset{H_{0}}{\gtrless}} \eta.$$



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- Let us define the whitened and normalized versions of $\mathbf{s}(\theta)$ and $\mathbf{r}:$

$$\begin{split} \mathbf{s}(\theta) &= \quad \frac{\Gamma^{-1/2} \, \mathbf{d}(\theta)}{\left(\mathbf{s}(\theta)^{H} \, \Gamma^{-1} \, \mathbf{s}(\theta)\right)} \\ \mathbf{u} &= \quad \frac{\Gamma^{-1/2} \, \mathbf{r}}{\left(\mathbf{r}^{H} \, \Gamma^{-1} \, \mathbf{r}\right)} \end{split}$$

• So that we can rewrite the NMF as:

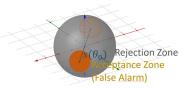
$$\mathbf{t}_{\Gamma}(\mathbf{r},\boldsymbol{\theta}) = \left| \mathbf{s}^{\mathsf{H}} \, \mathbf{u} \right|^{2} \underset{\mathsf{H}_{0}}{\overset{\mathsf{H}_{1}}{\gtrsim}} \eta. \tag{1}$$

with \mathbf{u} and $\mathbf{s}(\boldsymbol{\theta})$ on the unit complex N-sphere.



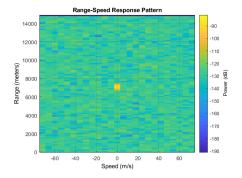
Mismatch and off-grid targets

- The NMF test quantity is the cosinus of the angle between s and u.
- In practice, θ is unknown. Mismatch $\delta = \theta - \theta_0$ with θ_0 the test parameter.
- Whatever the signal power α value, if the mismatch is too big the signal is not detected.





The Grid Principle



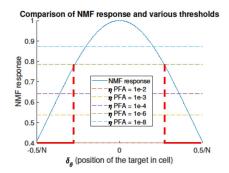
- The tests are run considering fixed values of θ in a grid G.
- In practice $\theta \neq \theta_0 \rightarrow$ mismatch \rightarrow drop in detection performance.

• Grid cells :
$$\left[\frac{k}{N} - \frac{1}{2N}, \frac{k}{N} + \frac{1}{2N}\right]$$
, parameters $\theta_0 = \frac{k}{N}$



Impact of off-grid targets on the NMF

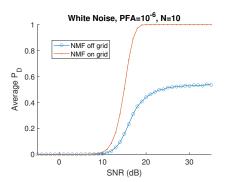
- The angle mismatch degrades the NMF response.
- θ uniformly distributed in a cell
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- The angle mismatch degrades the NMF response.
- θ uniformly distributed in a cell \rightarrow Expected Probability of Detection (P_D): <u>domain under threshold</u> whole domain
- P_D → 1 [Rabaste et al., 2016]
- It is even worse when $\Gamma \neq I$





Existing Solutions

• Extension of the GLRT to off-grid targets:

$$\mathsf{GLRT}(\mathbf{r},\theta_0) = \max_{\theta_c \in [\theta_0 - \Delta/2, \theta_0 + \Delta/2]} \mathsf{t}_{\Gamma}(\mathbf{r},\theta_c) \overset{\mathsf{H}_1}{\underset{\mathsf{H}_0}{\gtrsim}} \eta.$$

The best P_D , no closed form available, threshold unknown, precise approximation can be costly.



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- Existing sub-optimal cost-efficient solutions include
 - Oversampling approximate GLRT, threshold unknown
 - Using DPSS subspace to approximate the cell structure, analytical threshold [Bosse and Rabaste, 2018]
 - Using a monopulse-inspired detection scheme that approximates the GLRT, no analytical threshold [Develter et al., 2021].



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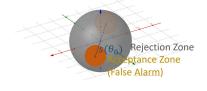
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- These solutions do not correct the convergence issue for all Γ and are not always near GLRT. This motivates studies on the true GLRT.



Finding the P_{FA} of the NMF geometrically

- Since **u** is whitened, it is distributed uniformly over the whole sphere.
- Thus $P_{FA} = \frac{Acceptance Zone Surface}{Rejection Zone Surface}$.
- This gives: $P_{FA} = (1 w^2)^{N-1}$.





• The GLRT corresponding to the NMF when θ is considered as an unknown parameter over a search domain \mathcal{D} gives:

$$\mathsf{GLRT}(\mathbf{r}, \mathcal{D}) = \max_{\boldsymbol{\theta}_{c} \in \mathcal{D}} \left| \mathbf{s}(\boldsymbol{\theta}_{c})^{\mathsf{H}} \mathbf{u} \right| \underset{\mathsf{H}_{0}}{\overset{\mathsf{H}_{1}}{\geq}} w'$$



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- The threshold is unknown, different from the NMF.
- It depends on the distribution of a continuum of non-independant variables → a priori hard to compute analytically.
- We use a geometrical method akin to the previous one.



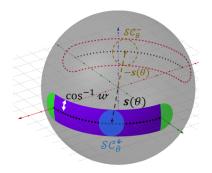
Methodology for real signals

- Consider for now \mathbf{u} and $\mathbf{s} \in \mathbb{R}^{N}.$
- **u** falls in the spherical cap \mathcal{SC}_{θ} corresponding to any $\theta \in \mathcal{D} \iff$ $\mathbf{u} \in \bigcup_{\theta \in \mathcal{D}} \mathcal{SC}_{\theta} \to A$ target is detected



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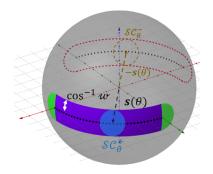
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- $\bigcup_{\theta \in \mathcal{D}} \mathcal{SC}_{\theta}$ forms a tube.
- Thanks to [Hotelling, 1939] we can compute the volume of this tube as the product of the cross section by the length of the curve.
- Valid as long as there is no overlap (the tube draws back into itself).





Rewriting our problem with real vectors

- Now \mathbf{u} and $\mathbf{s} \in \mathbb{C}^{N}$.
- $\left| \mathbf{s}(\theta)^{H} \mathbf{u} \right| = \max_{\alpha} \operatorname{Re} \left(\mathbf{s}(\theta)^{H} \mathbf{u} e^{-i\alpha} \right).$
- Translating our problem to real signals \rightarrow we have to maximize on the phase α of the product $\mathbf{s}(\theta)^H \mathbf{u}$.



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- Translating our problem to real signals \rightarrow we have to maximize on the phase α of the product $\mathbf{s}(\theta)^H \mathbf{u}$.
- In the end, we find that

$$\max_{\boldsymbol{\theta}_{c} \in \mathcal{D}} \left| \mathbf{s}(\boldsymbol{\theta}_{c})^{H} \mathbf{u} \right| = \max_{\{\boldsymbol{\theta}_{c}, \boldsymbol{\alpha}\} \in \mathcal{D} \times [0, 2\pi]} (\boldsymbol{\gamma}_{1}(\boldsymbol{\theta}_{c}) \cos \boldsymbol{\alpha} + \boldsymbol{\gamma}_{2}(\boldsymbol{\theta}_{c}) \sin \boldsymbol{\alpha})^{T} \underline{\mathbf{u}}$$

where
$$\gamma_1(\theta_c) = \begin{bmatrix} \mathbf{s}_r (\theta_c) \\ \mathbf{s}_i (\theta_c) \end{bmatrix}$$
, $\gamma_2(\theta_c) = \begin{bmatrix} -\mathbf{s}_i (\theta_c) \\ \mathbf{s}_r (\theta_c) \end{bmatrix}$ and $\underline{\mathbf{u}} = \begin{bmatrix} \mathbf{u}_r \\ \mathbf{u}_i \end{bmatrix}$ is a 2N-real valued noise vector drawn uniformly on S^{2N-1} under H_0 .



Methodology with complex signals

• We have to compute the volume of the tube around the 2D manifold $\gamma(\alpha, \theta) = \gamma_1(\theta) \cos \alpha + \gamma_2(\theta) \sin \alpha$.



Methodology with complex signals

- We have to compute the volume of the tube around the 2D manifold $\gamma(\alpha, \theta) = \gamma_1(\theta) \cos \alpha + \gamma_2(\theta) \sin \alpha$.
- Not covered by Hotelling, however [Johnstone and Siegmund, 1989] gives the following theorem:

Theorem

 $\begin{array}{l} \mbox{For $i \in [1,2]$, let $\gamma_1: [0,t_0] \rightarrow S^{n-1}$ be regular curves. Assume $\gamma_1(t)^T \gamma_2(t) = 0$ for all t. Let $Z(t) = \left[\left(\gamma_1(t)^T u\right)^2 + \left(\gamma_2(t)^T u\right)^2\right]^{1/2}$ where u is uniformly distributed on S^{n-1}. Then for $0 < w < 1$, we have: } \end{array}$

$$\begin{split} \mathbb{P} \left(\max_{0 \leq t \leq t_{0}} Z(t) > w \right) &\leq (1 - w^{2})^{(n-2)/2} + \frac{\Gamma\left(\frac{n}{2}\right) w (1 - w^{2})^{(n-3)/2}}{2\pi^{3/2} \Gamma\left(\frac{n-1}{2}\right)} \\ &\times \int_{0}^{t_{0}} \int_{0}^{2\pi} \left[\|\dot{\gamma}_{1}(t) \cos \omega + \dot{\gamma}_{2}(t) \sin \omega\|^{2} - \left(\dot{\gamma}_{1}(t)^{T} \gamma_{2}(t)\right)^{2} \right]^{1/2} d\omega \, dt \,, \end{split}$$

where $\dot{\gamma}_i(t)$ is the derivative of $\gamma_i(t)$ with respect to t. When there is no overlap, this inequality becomes an equality.



Final Results

Applying the previous theorem in our case, we get the following corollary :

Corollary

In the absence of overlap (low P_{FA} regimes), the P_{FA} for the NMF-GLRT for a search interval $\mathcal{D} = [\theta_1, \theta_2]$ with the steering vector $s(\theta)$ defined previously is given by:



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• Under white noise ($\Gamma = \sigma^2 \mathbf{I}$):

$$P_{FA} = \underbrace{(1 - w^2)^{N-1}}_{\prod \frac{1}{3}} + \underbrace{(1 - w^2)^{N-\frac{3}{2}}}_{\Gamma \left(N - \frac{1}{2}\right)} \left(N^2 - 1\right)^{\frac{1}{2}} \underbrace{(\theta_2 - \theta_1)}_{\text{length of search domain}}.$$
(2)



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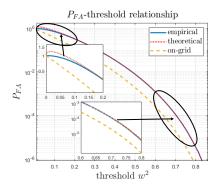
$$P_{FA} = \underbrace{(1 - w^2)^{N-1}}_{\prod \frac{1}{3}} + \underbrace{(1 - w^2)^{N-\frac{3}{2}}}_{\prod \frac{1}{3}} (N^2 - 1)^{\frac{1}{2}} \underbrace{(\theta_2 - \theta_1)}_{\text{length of search domain}}.$$
(2)

 Under colored noise (Γ ≠ σ² I), the integral in (2) can be evaluated numerically.



Numerical evaluation

- Plot of the previous relation with $\Gamma = I$ against simulated threshold with 10^8 noise samples, N = 10, $\mathcal{D} = [0, 1/N]$, GLRT approximation with 30 tests.
- Low P_{FA} of practical interest in Radar → the formula fits perfectly.
- Very High P_{FA} → slight mismatch because of overlap phenomena.





- In this work, we derived a P_{FA}-threshold relationship for the GLRT of the NMF extended to an unknown parameter of the target. It is exact for low P_{FA} of practical interest in Radar.
- Perspectives include the extension of this work to STAP (2D) detection and writing a journal paper deriving conditions under which no overlap arises.



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